Model 584 Constant-Fraction Discriminator Operating and Service Manual

Advanced Measurement Technology, Inc.

a/k/a/ ORTEC[®], a subsidiary of AMETEK[®], Inc.

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SAFETY INSTRUCTIONS AND SYMBOLS

This manual contains up to three levels of safety instructions that must be observed in order to avoid personal injury and/or damage to equipment or other property. These are:

- **DANGER** Indicates a hazard that could result in death or serious bodily harm if the safety instruction is not observed.
- **WARNING** Indicates a hazard that could result in bodily harm if the safety instruction is not observed.
- **CAUTION** Indicates a hazard that could result in property damage if the safety instruction is not observed.

Please read all safety instructions carefully and make sure you understand them fully before attempting to use this product.

In addition, the following symbol may appear on the product:





Please read all safety instructions carefully and make sure you understand them fully before attempting to use this product.

SAFETY WARNINGS AND CLEANING INSTRUCTIONS

DANGER Opening the cover of this instrument is likely to expose dangerous voltages. Disconnect the instrument from all voltage sources while it is being opened.

WARNING Using this instrument in a manner not specified by the manufacturer may impair the protection provided by the instrument.

Cleaning Instructions

To clean the instrument exterior:

- Unplug the instrument from the ac power supply.
- Remove loose dust on the outside of the instrument with a lint-free cloth.
- Remove remaining dirt with a lint-free cloth dampened in a general-purpose detergent and water solution. Do not use abrasive cleaners.

CAUTION To prevent moisture inside of the instrument during external cleaning, use only enough liquid to dampen the cloth or applicator.

• Allow the instrument to dry completely before reconnecting it to the power source.



3 æ UNGED GATED GATED INPUT LOBIC -IZV HEV KORMA HEV SCIOMA -EV SCIOMA

ORTEC MODEL 584 CONSTANT-FRACTION DISCRIMINATOR

1. DESCRIPTION

1.1. PURPOSE

The ORTEC 584 allows excellent time resolution from all commonly used detectors such as highpurity germanium (HPGe), surface barrier, fast plastic, Nal(TI), and photomultiplier tubes. This unit accepts input pulses in the range of 0 to -5 V and generates NIM fast negative outputs and a slow positive output that are based on the constantfraction time derivation technique. The 584 operates as an integral discriminator and provides excellent timing characteristics for a wide dynamic range of input signal amplitudes. Three timing modes are provided in the 584: constant fraction (CF), constant fraction with slow rise time reject (SRT), and leading edge (LE). Also, the 584 can be gated by the Bin Gate, a slow positive NIM signal, or a fast negative NIM signal.

The input impedance of the 584 is 50 Ω , and the NIM fast logic output signals are designed for termination in 50 Ω . A NIM slow positive SCA output is also furnished with an output impedance <10 Ω .

1.2. DISCRIMINATION CHARACTERISTICS

The 584 operates as an integral discriminator. Each input pulse that exceeds the lower-level threshold (adjustable on the front panel) can cause a set of timing output signals to be generated. This set of signals includes two independent fast negative NIM signals, nominally 5 ns in width and a slow positive NIM signal adjustable in width from $\leq 0.5 \ \mu s$ to $\geq 2.5 \ \mu s$.

The 584 includes a slow rise time reject function, selectable by the operator, and a variable width blocking one-shot. The slow rise time reject circuit can be used to inhibit the discriminator response to input signals that would cause leading edge timing. A blocking one-shot pulse is initiated simultaneously with the timing output signals. The blocking one-shot prohibits further timing output signals from being produced during the adjusted

blocking period. A negative NIM blocking output pulse is provided through a front panel connector. The width of the blocking output is determined by the width set for the blocking one-shot.

1.3. CONSTANT-FRACTION PRINCIPLE

In the constant-fraction technique, an input signal to the constant-fraction circuitry is delayed, and a fraction of the undelayed pulse is subtracted from it. A bipolar signal is generated, and its zero crossing is detected and used to produce an output logic pulse.

The constant-fraction shaping delay is controlled by the length of cable that is connected externally between the two Delay connectors on the front panel. The shaping delay should be optimized for each specific application. This optimization requires prior knowledge of the rise time and nominal width of the input signals to the 584.

1.4. INPUT/OUTPUT CHARACTERISTICS

The 584 accepts negative input signals to -5 V without saturation. If an input signal satisfies the logic conditions established with the 584, four output pulses are initiated simultaneously. Two timing output signals are provided through front panel connectors; these are independent negative NIM fast logic pulses nominally 5 ns in width. A positive NIM output is provided through a front panel BNC. The blocking output is a negative current pulse that is a NIM fast logic pulse; its width is determined by the front panel Width control setting. The blocking output is furnished through a front panel connector. The width of the blocking output is set by the period of the internal blocking one-shot, which is variable in the range of ≤ 10 ns to ≥1000 ns. The purpose of the blocking one-shot is to prohibit further timing output signals from being generated during the blocking period.

2. SPECIFICATIONS

2.1. PERFORMANCE

INPUT Accepts negative input signals from 0 V to -5 V without saturation; dc-coupled; $Z_{in} = 50 \Omega$; reflections \leq +5% for $t_r \geq 2$ ns.

THRESHOLD RANGE -5 mV to -1 V.

THRESHOLD INTEGRAL NONLINEARITY $\leq \pm 0.25\%$ of full scale.

PROPAGATION DELAY Nominally 25 ns, with external CF Delay ≅2 ns.

MINIMUM PULSE-PAIR RESOLUTION ≤20 ns.

DEADTIME Nominally 20 ns or blocking output width, whichever is greater.

BLOCKING OUTPUT WIDTH Adjustable from ≤ 10 ns to ≥ 1000 ns.

TIME WALK $\leq \pm 100$ ps for the 100:1 input range from -20 mV to -2 V; $\leq \pm 150$ ps for the 200:1 input range from -10 mV to -2 V. Conditions: external CF Delay \approx 2 ns; input rise time ≤ 1 ns; input pulse width \approx 10 ns.

2.2. CONTROLS

THRESHOLD Front panel 10-turn precision locking potentiometer determines the discriminator threshold setting in the range from -5 mV to -1 V.

TIMING MODE SWITCH Front panel 3-position locking toggle switch selects one of the three timing modes:

CF (Constant Fraction) Attenuation factor is internally set at f = 0.2 (can be changed upon request). An external 50 Ω coaxial cable must be provided for the constant- fraction shaping delay (CF Delay).

SRT(Slow Rise Time Reject) Provides constant - fraction timing and inhibits output signals that would be produced by leading-edge timing from the leading-edge arming discriminator. An input signal that does not cross the discriminator threshold before the constant-fraction zero-crossing time does not produce an output pulse.

LE (Leading Edge) Inhibits timing from the constant- fraction circuitry. The timing is derived as the leading edge of the input signal crosses the discriminator threshold level.

CF DELAY Two front panel BNC connectors accept 50 Ω coaxial cable to set the required constant-fraction shaping delay for the CF and SRT Modes: total delay is ≈ 0.8 ns plus the delay of the external cable. In the LE Mode, the user may either connect a piece of 50 Ω coaxial cable between these two connectors or connect a 50 Ω termination to each of the two connectors.

WALK Front panel 20-turn screwdriver adjustment sets the walk compensation for each application.

CF MON Front panel BNC connector permits observation of the constant-fraction bipolar timing signal; $Z_o = 50 \ \Omega$. 50 Ω coaxial cable required; 50 Ω termination suggested.

WIDTH Front panel 20-turn screwdriver adjustment sets the width of the blocking output pulse. Variable from ≤ 10 ns to ≥ 1000 ns. Sets the instrument deadtime for widths nominally >20 ns.

GATING MODE SWITCH Rear panel 2-position locking toggle switch controls the use of the Gate Inputs. (One of three Gate Input signal paths is selected by a printed wiring board, PWB, jumper.)

Gated A "true" logic level from the selected Gate Input permits output signals to be generated by the discriminator. A "false" logic level from the selected Gate Input inhibits output signals from being generated by the discriminator. A set of output signals already in progress is not terminated prematurely by a logic "false" signal from the selected Gate Input.

Ungated The signal level of the selected Gate input does not inhibit normal generation of output signals from the discriminator (i.e., the discriminator is always enabled).

LOGIC CURRENT SWITCH Rear panel 2-position locking toggle switch selects either the -6 V or the -12 V NIM supply line for providing current for the high-speed ECL logic used in the discriminator. NOTES:

(1) The module is within the current allotment for a single NIM width when using the -6V position with a NIM Class V Power Supply or equivalent.

(2) The module exceeds the current allotment for a single NIM width on the -12V supply when using the -12 V position. However, this position permits using the discriminator in bins with power supplies not providing -6 V.

GATE INPUT JUMPER (G+, G-, or BG) PWB jumper selects one, of three Gate Input signal paths.

G+ Selects the rear panel BNC Gate input connector to accept slow positive NIM input signal levels for gating; dc-coupled; $Z_{in} > 1 k\Omega$.

G- Selects the rear panel BNC Gate input connector to accept fast negative NIM input signal levels for gating; dc-coupled; $Z_{in} = 50 \Omega$.

BG Selects the Bin Gate line (pin 36 of the NIM power connector block) to accept slow positive NIM input signal levels >+2 V for gating; dc-coupled; Z_{in} > 1 k Ω .

POSITIVE OUTPUT WIDTH (+ Width) PWB 4turn potentiometer sets the width of the slow positive NIM output signal in the range from ≤ 0.5 , µs to ≥ 2.5 µs.

POSITIVE OUTPUT SIGNAL POLARITY (PO or PO) PWB jumper selects the slow positive NIM output signal (PO) or the complement output signal (PO).

2.3. INPUTS

INPUT Front panel BNC connector accepts fast negative input signals from 0 V to -5 V without saturation; dc-coupled; $Z_{in} = 50 \Omega$; reflections $\leq 5\%$ for $t_r \geq 2$ ns.

GATE INPUT Rear panel BNC connector; input signals accepted according to PWB Gate Input Jumper.

G+ Jumper Position Accepts slow positive NIM input signal levels for gating; dc-coupled; $Z_{in} > 1 \text{ k}\Omega$.

G-Jumper Position Accepts fast negative NIM input signal levels for gating; dc-coupled; $Z_{in} = 50 \ \Omega$.

2.4. OUTPUTS

TIMING Two front panel BNC connectors provide simultaneous NIM-standard fast negative logic signals; $t_r \cong 2 \text{ ns}$; $t_f \cong 3 \text{ ns}$; $t_w \cong 5 \text{ ns}$.

BK OUT Front panel BNC connector provides a NIM-standard fast negative logic pulse that occurs simultaneously with the timing outputs; width variable by front panel adjustment from \leq 10 ns to \geq 1000 ns; t_r \cong 2 ns.

POSITIVE Front panel BNC connector provides NIM-standard slow positive logic pulse simultaneously with timing outputs; $Z_o < 10 \Omega$; width variable by PWB width adjustment from $\le 0.5\mu$ s to $\ge 2.5\mu$ s. The associated LED is triggered for approximately 3 ms (updating) by each positive output pulse.

2.5. ELECTRICAL AND MECHANICAL

WEIGHT

Net 1.2 kg (2 lb 11 oz). **Shipping** 2.25 kg (5 lb).

DIMENSIONS Standard single-width NIM module (1.35 by 8.714 in.) per TID. 20893 (Rev).

POWER REQUIRED

Logic Current Switch*				
	Position			
	-6 v	-12 V		
	(mA)	<u>(mA)</u>		
+12 V	100	100		
-12 V	100	500		
+6 V	0	0		
-6 V	400	0		
+24 V	0	0		
-24 V	80	80		
<u>117 V ac</u>	0	0		

*See **"NOTES"** on Logic Current Switch, "Controls" Section of Specifications.

3. INSTALLATION

3.1. GENERAL

The 584 power requirements must be furnished from a NIM-standard bin and power supply such as the ORTEC 4001/4002 Series. The bin and power supply into which the 584 will normally be installed for operation is designed for rack-mounting. If the equipment is rack-mounted, there must be adequate ventilation to prevent any localized heating in the 584. The temperature of equipment mounted in racks can easily exceed the maxi mum limit of 50°C (323 K) unless precautions are taken.

3.2. CONNECTION TO POWER

Turn off the bin power supply before inserting or removing any modules. To be sure of proper operation, check the dc voltage levels of the power supply after all modules have been installed in the bin. ORTEC bins and power supplies include convenient monitoring test points on the power supply control panel.

3.3. INPUT CONNECTION

The 584 Input connector on the front panel accepts negative input pulses. This input is terminated internally in 50 Ω . Connect the source of negative input signals to this connector through a 50 Ω coaxial cable.

3.4. OUTPUT CONNECTIONS

The primary outputs of the 584 are the NIMstandard fast negative logic signals that are provided through the front panel BNC connectors marked Timing Outputs. The circuits driving these two connectors are independent so that the Timing Outputs can be used independently. Each negative NIM output that is used must be terminated in 50 Ω , and 50 Ω cable must be used for the interconnections. Unused outputs need no termination.

The Positive Output on the front panel provides a NIM-standard slow positive pulse that is initiated simultaneously with the Timing Outputs on the front panel. This output can be used with instruments such as counters or ratemeters which require

positive input signals. The interconnection can be made with 93 Ω cable, terminated in 100 Ω or more, in most applications.

3.5 CONSTANT-FRACTION DELAY CABLE

The constant-fraction time derivation circuit is not complete until an external length of 50 Ω cable has been connected between the two Delay connectors on the front panel. The total constant-fraction shaping delay is equal to the external constant-fraction shaping delay, $t_{\rm d(Ext)}$, plus approximately 0.8 ns.

One important use of the 584 is in fast timing or counting experiments with scintillators and photomultiplier tubes (PMTs). In these applications, the total constant-fraction shaping delay, $t_{d(Tot)}$, is selected so that the zero crossing of the bipolar timing signal occurs after the attenuated, undelayed portion of the constant-fraction signal has reached its maximum amplitude. Thus the zero crossing occurs at the same fraction of the input pulse height regardless of the amplitude of the input signal. Selection of the constant-fraction shaping delay for best timing performance with a given scintillator and PMT is usually accomplished experimentally. A useful empirical formula for the initial trial selection of the external shaping delay is

 $t_{d(Ext)} \cong 1.1 t_r - 0.8 ns,$

where t_r is the 10% to 90% rise time of the anode pulses. Walk adjustment can then be accomplished while observing the delayed CF Mon signal on a fast oscilloscope which is triggered externally by a timing output signal from the 584.

An alternate, more accurate method is to make repeated measurements of the timing spectra of a timing coincidence system: first, for small (10%) changes in the initial selection of the external shaping delay, and second, for small changes in the Walk adjustment. Another important use of the 584 is timing with HPGe detectors. In this application, the constantfraction shaping delay selected is less than the minimum signal rise time. Depending on the size and shape of the HPGe detector, the rise time can range from 50 to 400 ns. A typical optimal constantfraction shaping delay can range from 10 to 30 ns with 20 ns a reasonable first choice. An ORTEC 425A Nanosecond Delay can be used to adjust the shaping delay. Timing with HPGe detectors, especially those with large volume, involves several interactive parameters. Refer to Section 6, Typical Applications, and refs. 1 and 2 in Section 6.5.

Other important uses of the 584 involve surface barrier detectors, Nal (TI) scintillators, microchannel plates, laser diodes, and other time pick-off applications (Section 6). Contact your local ORTEC representative for the latest information on state-of-the-art timing applications.

4. OPERATING INSTRUCTIONS

The 584 accepts negative input signals in the range of 0 to -5 V. For each input pulse that satisfies the 584 logic criteria, four output logic pulses are initiated simultaneously. Two timing output signals are provided: these are negative NIM fast logic pulses that are nominally 5 ns wide. A positive slow NIM output is provided through a front panel connector. The front panel Blocking Output is a negative current logic signal similar to a NIM fast logic pulse but with a width that is set by the period of the internal blocking one-shot, and this is variable from <10 ns to \ge 1000 ns. An SRT timing mode can be employed to ensure that leading edge time walk is not introduced by the leading edge discriminator. Leading edge timing can occur for input signals with exceptionally long rise times and for amplitude-and-rise time compensated (ARC) timing with input signals that exceed the threshold of interest by only a slight amount. In the SRT mode, an input signal that does not cross the lower level threshold before the constant fraction zero crossing time will not produce output signals.

5. CIRCUIT DESCRIPTION

5.1. OVERVIEW

Figure 5.1 is a simplified block diagram of the ORTEC 584 and can be used as a reference to describe its operation.

The 584 can be operated in three separate timing modes that are selectable by a front panel locking toggle switch. These three timing modes are Constant Fraction (CF), Slow Rise Time Reject (SRT), and Leading Edge (LE). The initial circuit description will only cover CF operation. All logic functions are implemented using ECL components

A 0 to -5 V input starts at time zero and is applied to the leading edge discriminator, LE. The

reference level for LE is set by the front panel Threshold control and can range from -5 mV to -1 V. The reference level is buffered from LE by A1.

The input signal is attenuated by RI and R2 and delayed by the external CF Shaping Delay. The attenuation factor is approximately f = 0.2. The attenuated and delayed input signals are differentially summed in the constant-fraction amplifier, CFA. The resulting bipolar shaped signal is fed to the zero crossing gate, G1. Note that the CFA signal can produce an output from G1 only if G1 is armed by the output signal from LE.

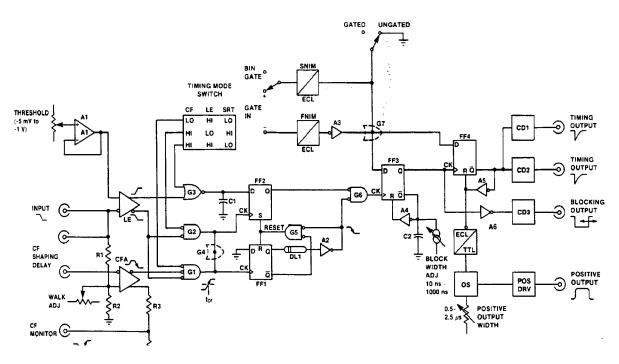


Fig. 5.1. Simplified Block Diagram of the ORTEC 584 Constant Fraction Discriminator.

The CF signal crosses the logic threshold of G1 at constant-fraction time, t_{cf}. Constant-fraction time is determined by the selection of the external constant-fraction shaping delay, t_{d(Ext)}, subject to the rise times and pulse shapes of the input signals. A fixed internal delay of about 0.8 ns must be added to $t_{d(Ext)}$ to determine the total constant-fraction shaping delay, $t_{d(Tot)}$. Note that the time of occurrence of the output of G1 is directly related to t_{cf} only if the arming signal from LE arrives at the input of G1 prior to the negative-going excursion of the output of CFA. If the signal from LE arrives at G1 later in time, the output signal from G1 will have time movement relative to the input signal. This type of time movement is often referred to as leading edge time walk.

Access to the inverted CF signal is provided by the front panel CF Mon connector. This signal is buffered by R3 and has a typical amplitude of 40 mV peak-to-peak when viewed with a 50 ohm input impedance oscilloscope. The baseline of the CF Mon signal has a -60 mV dc off set into 50 ohm, and a -120 mV offset open circuit.

The G1 output signal clocks the high state at D to the Q output of FF1. The Q output of FF1 is delayed by DL1, inverted by A2, and applied to the input of G6. The D input of FF2 is in the low state since the timing mode switch in the CF position applies a high state to the input of G3.

Note that a previous reset signal preset the Q output of FF2 to a high state. When the G1 output signal clocks FF2, through G4, the D input low state is clocked to the Q output which enables G6 to pass the timing signal from A2. The output of A2 initiates reset for FF1 and FF2 through G5. The Q output of FF2 is reset to a high state which produces a pulse <10 ns wide at the output of G6.

The output signal from G6 is processed by the output shaping circuitry. Assuming that the rear panel switch is in the Ungated position, a high state is present on the D inputs of FF3 and FF4. The output of G6 clocks FF3 causing the Q output to change states which, in turn, clocks FF4, causing the Q output of FF4 to change states. Flip-flop FF4 is reset by A5 to produce a nominally 5-ns wide

pulse. This clipped pulse is furnished through current drivers CD1 and CD2 to the timing outputs, and to an inverting ECL-to-TTL converter to produce the positive output. The width of the positive output is adjustable from $\leq 0.5 \ \mu s$ to $\geq 2.5 \ \mu s$ by a printed-wiring board mounted potentiometer. Additional output signals are blocked by the blocking one-shot composed of FF3, A4, C2, and the Blocking Width control. The blocking period is determined by the adjustable delay time to reset FF3. The blocking period is variable from $\leq 10 \ ns$ to $\geq 1000 \ ns$ and can be monitored at the front panel connector.

When operated in the Leading Edge timing mode (LE), the timing mode switch disables G1 and G3 while enabling G2. The circuit operation is similar to the CF mode except that the LE discriminator output, through G2, clocks FF1 and FF2 to produce all the output signals. The timing output signals are no longer related in time to the zero crossing time of the constant-fraction signal. Rather, all output signals are related to the time the leading edge of the input signal crosses the LE threshold level.

When operated in the Slow Rise Time Reject mode (SRT), the timing mode switch disables G2 and enables G1 and G3. Operation differs from the CF only by the function of FF2. The inverted output of the LE discriminator gates G3 causing its output to switch to a low state on a time constant set by C1. If the output of G3 decays below the threshold level of FF2 prior to t_{cf} , the unit functions the same

as in the CF mode. However, if t_{cf} occurs prior to a signal at the output of G3, the Q output, FF2, will not change state when clocked by the output of G1. Thus all output signals will be blocked at G6.

Four separate gating modes are available. A rear panel locking toggle switch selects either Gated or Ungated operation. When set in the Ungated position, a high state is connected to the D inputs of FF3 and FF4, which enables output signals to be generated. When set in the Gated position, a printed wiring-board jumper must be set to select the Bin Gate (pin 36 of the NIM power connector block), a slow positive NIM input via a rear panel connector, or a fast negative NIM signal via the same rear panel connector. The slow positive NIM signal and the Bin Gate are converted to ECL signal levels by a NIM-to-ECL converter. The fast negative NIM signal is converted to ECL signal levels by a fast NIM-to-ECL converter and A3. All gating options are wired at G7.

The input power requirements are +12 V, -12 V, -6 V, and -24 V. A rear panel switch allows use of the -12 V supply instead of the -6 V supply to provide the ECL logic current. These levels are all obtained from the bin power supply. Additional voltage levels of +5 V and -5.2 V are also required. These voltages are obtained from regulators on the printed wiring board.

6. APPLICATIONS

6.1. TIMING WITH FAST SCINTILATORS

Figure 6.1. shows a typical system for timing with fast scintillation detectors such as Naton-136, Pilot B, KL-236, NE-102, NE-111, NE-213, etc. A 584 Constant-Fraction Discriminator is used in each of the two channels of the Time-to-Amplitude Converter (TAC). Selection of the constant-fraction shaping delay for best timing performance with a given scintillator and PMT is usually accomplished experimentally (Section 3.5). A useful empirical formula for the initial trial selection of the external shaping delay is

 $t_{d(Ext)} \cong$ 1.1 t_r – 0.8 ns

where t_r is the 10% to 90% rise time of the anode pulses.

The timing resolution obtainable in a given application with fast scintillators depends on many variables, including the energy range selected by the SCAs; the threshold setting, external CF Delay, and walk adjustment of the 584; the rise time and type of the PMT; the size and shape of the scintillator; and the presence of noise in the system. Typical timing resolution for a narrow dynamic range (.1:1.1), for ⁶⁰Co is \geq 200 ps FWHM and \geq 360 ps FWTM (Fig. 6.1.).

For additional data see Section 6.5, "References." Also, contact your local ORTEC representative to obtain the latest information on the state-of-the-art in timing with fast scintillators.

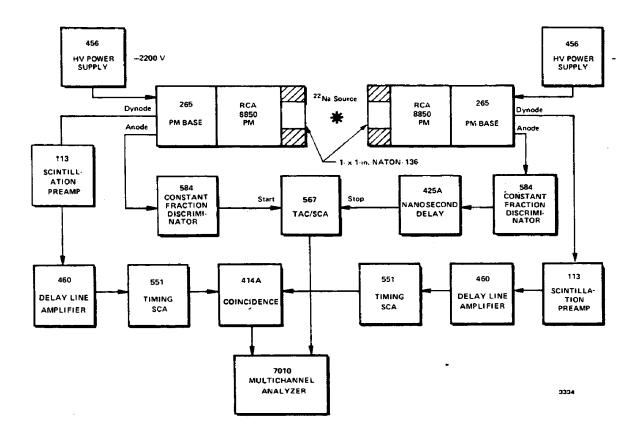


Figure 6.1. A System for Gamma-Gamma Lifetime Measurement.

6.2. TIMING WITH NaI(TI) SCINTILLATORS

Timing with Nal(TI)) scintillators is similar to timing measurements with fast scintillators. However, one additional problem must be considered. The photoelectron statistics for low-energy gamma-ray applications are so poor that individual events near the trailing edge of Nal(TI) pulses can trigger the 584. Thus a single scintillation event can produce two or more discriminator output pulses. In the 584, this problem is overcome by setting the Blocking Width adjustment to \leq 1000 ns. The 584 can also be used successfully on longer decay scintillators, but the maximum blocking width time may have to be increased to prevent multiple triggering.

The timing resolution obtainable with an Na I (TI) in a given application depends on many variables. Typical timing resolution for a properly adjusted system using a 1 in. x 1 in. KL-236 and an RCA 8575 PMT in the start channel, a 1 in. x 1 in. Nal(TI) and an RCA 8575 PMT in the stop channel, and a 50:1 dynamic range with 60 Co is \leq 900 ps FWHM and \leq 1.9 ns FWTM.

6.3. TIMING WITH LARGE VOLUME HPGE DETECTORS

Figure 6.2. shows a typical timing coincidence system using a large volume HPGe detector in the stop channel. In this application, the timing resolution is dominated by the charge collection characteristics of the HPGe detector. The output signal from the HPGe detector preamplifier must be shaped prior to processing by the 584. Either a Fast Filter Amplifier (ORTEC Model 579) or a Timing Filter Amplifier (ORTEC Model 474) can perform the necessary pulse shaping. In typical applications, the 579 is set at \geq 200 ns differentiation time constant and the integral time constant is set at Out.

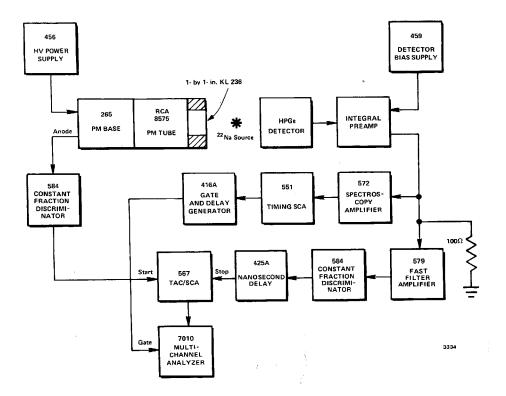


Figure 6.2. Gamma-Gamma Coincidence System Using a Plastic Scintillator and a Large HPGe Coaxial Detector

The slow rise time reject mode is particularly useful with HPGe detectors operated over a wide dynamic range of energies. Operation in the SRT mode will have minimum effect on the FWHM resolution but can dramatically improve the FWTM and FW(1/100)M values. The disadvantage of the SRT

mode is that the effective relative efficiency of the HPGe detector for timing is reduced since some events of valid energy are removed from the timing spectra.

The timing resolution obtainable with HPGe detectors also depends on many variables. The active volume of the HPGe detector is a major factor in determining timing resolution.

The charge collection time increases as the detector volume increases. Charge collection time variations result in variations in the shape and rise time of the preamplifier output signal. For this reason, Amplitude and Rise Time Compensated (ARC) timing is used in HPGe detector applications. The principal difference between ARC timing and CF timing is the selection of the CF shaping delay. In ARC timing, the CF shaping delay selected is less than the minimum rise time of the 579 output signal.

The optimum CF shaping delay for an HPGe detector is usually obtained experimentally. An ORTEC 425A Nanosecond Delay unit can be used to vary the shaping delay of the 584 while taking a series of timing spectra. A typical value for the shaping delay for a 15% HPGe detector is in the range of 20 ns to 30 ns.

Figure 6.3. shows the timing resolution FWHM for 14 HPGe detectors ranging in size from ~10% to 35% relative efficiency. Note that individual detectors can deviate significantly from the mean value of timing resolution.

6.4. TIMING WITH OTHER DETECTORS

The 584 can be used to provide timing information for other detectors such as ORTEC Surface Barrier Detectors and Low-Energy Photon Detectors. When input signals are low level and have very fast rise times, an ORTEC 9301 or 9305 Fast Preamplifier, or an ORTEC 574 or 535 Quad Fast Amplifier are recommended as accessory modules. ORTEC conducts a continuing program aimed at improving timing spectroscopy techniques. Please contact your local ORTEC representative concerning any special requirements.

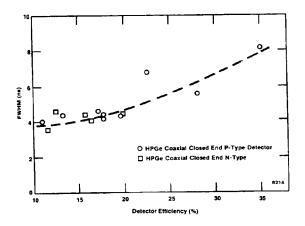


Fig. 6.3. Timing Resolution FWHM for 14 Detector Systems as a Function of Efficiency for the Energy Range 511 ± 50 keV for -- Na.

6.5. REFERENCES

Use the following references for further information on typical timing coincidence measurement systems in which the ORTEC 584 Constant-Fraction Discriminator can be used.

1. "Principles and Applications of Timing Spectroscopy," *Application Note AN 42,* ORTEC, Oak Ridge, TN (1982).

2. T. J. Paulus, T. W. Raudorf, B. Coyne, and R. C. Trammell, "Comparative Timing Performance of Large Volume HPGe Germanium Detectors," *IEEE Trans. Nucl. Sci.*, **NS-28**, No. 1, pp. 544-548 (1981).

3. M. 0. Bedwell and T. J. Paulus, "A Constant-Fraction Differential Discriminator for Use in Fast Timing Coincidence Systems," *IEEE Trans. Nucl. Sci.*, **NS-26**, No. 1, p. 442 (1979).

4. M.O.Bedwell and T.J.Paulus, "A Versatile Constant-Fraction 100-MHZ Discriminator," *IEEE Trans. Nucl. Sci.*, **NS-25**, No. 1, p. 86 (1978).

5. M. 0. Bedwell and T. J. Paulus, "A New Constant-Fraction Timing System with Improved Time Derivation Characteristics," *IEEE Trans. Null Sci.*, **NS-23**, p. 234 (1976).

6. G. F. Knoll, *Radiation Detection and Measurement,* John Wiley and Sons, New York, NY (1979).

This instrument can be returned to the ORTEC factory for service and repair at a nominal cost. The ORTEC standard procedure for repair ensures that the same quality control and checkout procedures that are used for a new instrument will be used for the repaired unit. Always contact Customer Services at ORTEC, (865) 482-4411, before

sending in an instrument for repair to obtain shipping instructions and so that the required Return Authorization Number can be assigned to the unit. This number should be written on the address label and on the package.

Bin/Module Connector Pin Assignments For Standard Nuclear Instrument Modules per DOE/ER-0457T.

Pin	Function	Pin	Function
1	+3 V	23	Reserved
2	- 3 V	24	Reserved
3	Spare bus	25	Reserved
4	Reserved bus	26	Spare
5	Coaxial	27	Spare
6	Coaxial	*28	+24 V
7	Coaxial	*29	- 24 V
8	200 V dc	30	Spare bus
9	Spare	31	Spare
*10	+6 V	32	Spare
*11	- 6 V	*33	117 V ac (hot)
12	Reserved bus	*34	Power return ground
13	Spare	35	Reset (Scaler)
14	Spare	36	Gate
15	Reserved	37	Reset (Auxiliary)
*16	+12 V	38	Coaxial
*17	- 12 V	39	Coaxial
18	Spare bus	40	Coaxial
19	Reserved bus	*41	117 V ac (neutral)
20	Spare	*42	High-quality ground
21	Spare	G	Ground guide pin
22	Reserved		

